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An improved mean-square weight deviation-proportionate gain algorithm based on error autocorrelation



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ABSTRACT

This paper presents an alternative approach to the gain distribution policy used in the z^2 -proportionate algorithm. The gain policy of the z^2 -proportionate uses a rule that combines the mean-square weight deviation-proportionate gain and a uniform one to obtain the whole algorithm gain distribution, leading to very good convergence characteristics. However, such a gain combination law is dependent on the knowledge of the measurement noise variance in the system, which in practice is not always readily available. Here, aiming to circumvent such dependence, a new strategy of gain distribution based on error autocorrelation is introduced. The proposed approach makes the use of the mean-square weight deviation-proportionate gain more attractive for real-world applications. Simulation results show that the proposed algorithm outperforms the z^2 -proportionate in terms of convergence characteristics for cases in which the measurement noise variance is either unknown or poorly estimated.

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1. Introduction

In many real-world applications, such as echo cancellation, channel equalization, and seismic processes, the plant impulse response is sparse [1–5]. For this class of responses, the well-known least-mean-square (LMS) and normalized LMS (NLMS) algorithms (which use the same adaptation step size for all filter coefficients) exhibit poor convergence characteristics [6–9]. To exploit the sparse nature of the plant impulse response, Duttweiler in [10] has proposed an algorithm, called proportionate NLMS (PNLMS), in which each filter coefficient is updated proportionally to its magnitude, resulting in an algorithm with improved convergence

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characteristics. However, the PNLMS fast initial convergence is not preserved over the whole adaptation process [9,11]. Furthermore, the PNLMS algorithm provides slow convergence when the plant exhibits medium and low sparseness [6,7]. To circumvent these problems, several versions of the PNLMS algorithm have been proposed in the open literature [7.9.12–14]. For instance, the improved PNLMS (IPNLMS) algorithm allows controlling its performance for convergence speed according to the plant sparseness, resulting in better convergence characteristics than the PNLMS for a wide range of plant sparseness [7]. Two versions of the PNLMS algorithm, termed sparseness-controlled PNLMS (SC-PNLMS) and sparseness-controlled IPNLMS (SC-IPNLMS), which take into account the sparseness variation of the plant are presented in [15] and [16], respectively; these algorithms perform well for both very high and medium sparseness levels, however at the expense of a high computational burden. Aiming to provide fast convergence during the whole adaptation



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process, μ -law PNLMS (MPNLMS) and adaptive MPNLMS (AMPNLMS) algorithms are given, respectively, in [11] and [17]; however, such algorithms are computationally much more complex than the PNLMS.

In [17,18], a new proportionate-type algorithm is presented. Such an algorithm, termed water-filling, is based on minimizing the mean-square error with respect to the algorithm gains for white input data. Also in [18], seeking to reduce computational complexity, the z^2 -proportionate algorithm is derived from the water-filling one. To deal with colored input data, a colored-water-filling (CWF) algorithm is discussed in [19,20]. This algorithm exhibits very good convergence characteristics: however, its computational load is too high. To circumvent such a drawback, two suboptimal versions of the CWF algorithm [suboptimal gain allocation version 1 (Suboptimal 1) and suboptimal gain allocation version 2 (Suboptimal 2)] are given in [19]. In contrast to the PNLMS-type algorithms that take into account the magnitude of the coefficients to compute the gain allocation, the algorithms discussed in [17-20] use an estimate of the mean-square weight deviation. Among the latter, the z^2 proportionate algorithm is the one that exhibits lower computational complexity and, for this reason, we concentrate here our study on this algorithm.

The z^2 -proportionate uses a gain distribution that combines the mean-square weight deviation-proportionate gain and a uniform one, such that the algorithm gain in steady state is uniform. This mixture of gains is controlled by a parameter that depends on the measurement noise variance, requiring therefore its estimate. Moreover, the z^2 -proportionate algorithm performance is degraded when such a variance value is not accurately estimated.

In this research work, a novel approach to obtain the algorithm gain distribution is proposed aiming to circumvent the dependence on the knowledge of the measurement noise variance. Thus, a new strategy to carry out the migration from the mean-square weight deviation-proportionate gain to uniform one is here introduced. Such a strategy, which does not require the knowledge of the measurement noise variance, is based on the autocorrelation between adjacent samples of the error signal. The proposed algorithm outperforms the z^2 -proportionate algorithm in terms of convergence characteristics for cases in which the measurement noise variance is either unknown or poorly estimated. Through numerical simulations, using a real-world echo path as plant impulse response, the effectiveness of the proposed algorithm is assessed for different operating scenarios.

This paper is organized as follows. Section 2 revisits the z^2 -proportionate algorithm. In Section 3, the impact of the measurement noise variance estimate error on the z^2 -proportionate algorithm behavior is briefly discussed. Section 4 presents the proposed algorithm as well as describes in detail its operation principle. In Section 5, numerical simulation results attest the performance of the new algorithm. Finally, Section 6 presents concluding remarks.

2. Revisiting the z^2 -proportionate algorithm

In this section, we start reviewing the general expressions of the PNLMS-type algorithms and next a brief description of the z^2 -proportionate algorithm is presented.

2.1. General expressions of the PNLMS-type algorithms

The PNLMS-type adaptive algorithms are formulated by the following set of equations [7,9]:

Coefficient update vector

$$\mathbf{w}(k+1) = \mathbf{w}(k) + \frac{\beta \mathbf{G}(k)\mathbf{x}(k)e(k)}{\mathbf{x}^{\mathsf{T}}(k)\mathbf{G}(k)\mathbf{x}(k) + \xi}$$
(1)

Error signal

$$e(k) = d(k) + v(k) - y(k)$$
 (2)

Adaptive filter output

$$y(k) = \mathbf{w}^{\mathrm{T}}(k)\mathbf{x}(k) = \mathbf{x}^{\mathrm{T}}(k)\mathbf{w}(k)$$
(3)

Gain distribution matrix

$$\mathbf{G}(k) = \operatorname{diag}[g_1(k) \, g_2(k) \cdots g_L(k)] \tag{4}$$

where $\mathbf{w}(k) = \begin{bmatrix} w_1(k) & w_2(k) & \cdots & w_L(k) \end{bmatrix}^T$ is the *L*-dimensional adaptive coefficient vector, $\mathbf{x}(k) = [x(k) x(k-1) \cdots x(k-L+1)]$ denotes the input vector with variance σ_x^2 , d(k) represents the desired signal, v(k) denotes a zero-mean independent and identically distributed (i.i.d.) measurement noise with variance σ_{ν}^2 and uncorrelated with any other signal in the system, and β is the step-size control parameter. Constant $\xi > 0$ is a regularization parameter that prevents division by zero in (1) and stabilizes the solution. The diagonal matrix G(k) distributes individual gains among the filter coefficients according to a specific rule. It should be noted that by setting the matrix $\mathbf{G}(k)$ equal to the identity matrix, the proportional adaptive algorithm becomes the NLMS algorithm. For the sake of clarity, we illustrate in Fig. 1 the block diagram of a typical adaptive filtering scheme showing the variables used here.

2.2. Brief description of the z^2 -proportionate algorithm

Let us define the weight deviation vector as

$$\mathbf{z}(k) = \mathbf{w}^{\mathbf{o}} - \mathbf{w}(k) \tag{5}$$

where $\mathbf{w}^{o} = [w_{1}^{o} w_{2}^{o} \cdots w_{L}^{o}]^{T}$ is an *L*-dimensional vector representing the plant impulse response. Thus, the gain of the *z*²-proportionate algorithm, obtained from the simplification of the water-filling algorithm, is given by [18]

$$g_i^{0}(k) = \frac{E[z_i^2(k)]}{(1/L)\sum_{j=1}^{L} E[z_j^2(k)]}, \qquad i = 1, 2, ..., L$$
(6)

with

$$E[z_i^2(k)] = E\{[w_i^0 - w_i(k)]^2\}$$
(7)

where $E[z_i^2(k)]$ represents the mean-square weight deviation of the *i*th coefficient.



Fig. 1. Typical setup of adaptive filtering.

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