

Contents lists available at [ScienceDirect](http://www.sciencedirect.com/science/journal/14348411)

Int. J. Electron. Commun. (AEÜ)

journal homepage: www.elsevier.com/locate/aeue

Short communication

Chirping techniques for enhancing the performance of SOA-based UWB over fiber systems: An experimental demonstration

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ARTICLE INFO

Keywords: Ultra wideband (UWB) Radio-over-fiber (RoF) Chirping Semiconductor Optical Amplifier (SOA) Nonlinearity Amplified Spontaneous Emission (ASE)

ABSTRACT

A phaser-based processing technique was adopted in an UWB over fiber system employing SOA. The target is to simultaneously reduce the ASE noise impact and nonlinear effects inherent to optical amplification. Experimental results prove the effectiveness of chirping in terms of cross correlation and bit error rate.

1. Introduction

The increasing demand for high speed and low power transmission systems nowadays, increases the attention towards ultra wideband (UWB) radio technology, which is a promising solution for new generations of short-range broadband wireless communications [\[1](#page--1-0)–3]. UWB was born in 2002, when the Federal Communication Commission (FCC) approved its free space propagation within a 7.5 GHz band and −41.3 dBm/MHz maximum emitted power [\[1\].](#page--1-0) UWB over fiber has known a growing interest over the past decade, as by exploiting the benefits of optical link a high quality of transmission may be achieved at long distances [\[4\]](#page--1-1). A particular scenario of UWB is the Impulse Radio, which is a cost effective approach offering a low order of generation complexity [\[5\]](#page--1-2) together with a simple design at the receiver side if a non-coherent detection is adopted $[6]$. With the view to get a reach extension in the optical access network, a low cost Semiconductor Optical Amplifier (SOA) has been recently employed [\[7,8\].](#page--1-4) However, a performance degradation may occur due to the nonlinear effects and Amplified Spontaneous Emission (ASE) noise inherent to SOA. In [\[8\]](#page--1-5), we have overcome these two impairments via phaser-based processing techniques [\[9\]](#page--1-6), a significant improvement in the power efficiency and bit error rate has been carried out, due to symmetrical up and down chirping performed at the transmitter and receiver respectively. The key concept is that a lower order of four wave mixing and cross gain modulation is obtained against signals with less frequency diversity. A

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<https://doi.org/10.1016/j.aeue.2018.07.008> Received 7 December 2017; Accepted 11 July 2018 1434-8411/ © 2018 Elsevier GmbH. All rights reserved. special effort is applied here to experimentally demonstrate the effectiveness of chirping in a practically implemented UWB over fiber system. Considering On Off Keying (OOK) and Pulse Position Modulation (PPM) formats, a better signal quality is achieved, and a lower number of errors is collected at receiver side, thanks to chirping.

2. UWB over fiber system architecture and experimental setup

[Fig. 1](#page-1-0) describes our basic radio over fiber system, the transmitter relies on a Mach–Zehnder Modulator (MZM) with half-wave voltage *Vπ* of 6 V, biased at $V_{DC} = 1.5 V_{\pi}$ by the electrical UWB signal, and with a continuous wave light at its input. The channel is made of a Single Mode Fiber (SMF) with a use of SOA for pre-amplification purpose. There is a growing interest for designing preamplified optical integrated receivers, and the SOA is a pertinent solution for such applications due to its low cost and complexity $[10,11]$. At the receiver side, a photodetector converts the optical power into an electrical voltage signal, the latter being attenuated before entering the antenna so as to meet FCC mask. The experimental setup is stated in [Fig. 2](#page-1-1), the transmitter relies on the Agilent Arbitrary Waveform Generator (AWG - M8190A), which electrically generates the signals designed in MATLAB, where the sampling frequency is 12 GHz and the maximum produced voltage is 700 mV. A radio frequency amplifier (SHF Communication Technologies AG - 115 BP) has been placed so as to boost the driving voltage entering the single arm MZM (Photline Technologies). On the optical

Fig. 1. Block diagram of the basic radio over fiber system.

source side, the laser diode has a fixed power of 12 dBm, which can be tuned via Attenuator (JDS Uniphase) before being injected to MZM. Then, the optical field propagates over 40 km of single mode fiber with a 0.2 dB/km attenuation and a dispersion coefficient of 17 ps/nm/km, to finally reach the SOA (Inphenix-IPSAD1581). The central wavelength of the optical filter (XTM-50) after SOA is 1540 nm, which corresponds to that at the laser output, and the 10 dB bandwidth is 30 pm. A periodic observation for the spectrum and average power is mandatory in order to re-calibrate the polarization controller after any systematical change in the fiber characteristics, as the latter are highly affected by the physical conditions of the surrounding environment; hence, an Optical Spectrum Analyzer (OSA - ADVANTEST Q8384) has been placed for that purpose. The photo-detector (HEWLETT PACKARD 11982A), with 15 GHz bandwidth, converts the optical field into an electrical voltage signal to be ready for entering the oscilloscope. Finally, the received waveform is sampled at Fs = 40 GHz, then saved for post processing in MATLAB.

3. Performance evaluation of phaser-based UWB over fiber system

The block diagram of the proposed system is described in [Fig. 3](#page--1-8), with up and down chirping taking place before and after the optical system respectively. The chirping-based transmission concept was initially explored by Nikfal [\[9\],](#page--1-6) who has succeeded in enhancing the signal-to-noise ratio of the impulse radio transceiver using phasers. The latter have dispersive structures with a controllable group delay [\[12\]](#page--1-9), which apply linear up or down chirping to stretch or compress the input signal without changing the total energy. So if the pulse is elongated by a time spread of M, the average power will be reduced by the same factor M. The transmitter operates a linear up chirping for the data signals before entering the channel, while down chirping with an absolutely equal slope is performed at receiver side, so the instantaneous frequencies are equalized and the signal is compressed to an enhanced waveform that exactly matches the original [\[9\].](#page--1-6) The time and frequency domain expressions of the chirped signal are described in Eqs. [\(1\)](#page-1-2)–(3), besides to the frequency dependent time delay function $\tau(f)$. The original pulse is $x(t)$, *a* is the chirping slope (which is positive for up chirping and negative for down chirping), where b is the delay offset. Eq. [\(3\)](#page-1-3) informs that chirping affects only the phase of the spectrum

without varying its magnitude, hence it has no influence on the power spectral density.

$$
y(t) = x(t + \tau(f))
$$
\n(1)

$$
\tau(f) = af + b \tag{2}
$$

$$
Y(f) = X(f) \exp(2\pi f \tau(f)) = X(f) \exp(2\pi af^2 j) \exp(2\pi f b j)
$$
 (3)

Chirping scheme has shown to be effective in our previous simulations, a very good power efficiencies being obtained for the 5th Gaussian pulse besides to Abraha's combination of Doublets [\[8\].](#page--1-5) Here we examine the potential of chirping in a real SOA-based optical system so as to confirm the effectiveness of this scheme, while up/down chirping is performed in offline processing via simple phaser models in MATLAB. Due to the band limitation of the AWG used, it was not possible to generate the same waveforms as in $[8]$, on the other side, dealing with pulses having a frequency diversity is much easier to see the benefits of chirping. Therefore, we have proposed a combination of 2 modulated basic Gaussians g_1 and g_2 , with different carrier frequencies f_1 and f_2 , in order to see the original signal as two overlapped waveforms which could be separated via chirping. It is preferred to absolutely separate g_1 from g_2 , in order to get a better readability and more accurate evaluation while comparing with the non-chirped signal, that is provided when satisfying the condition $f_2 \ge f_1$. The time duration of the utilized combination is 13.25 ns, where the pulse shaping factor for both Gaussians is $\sigma = 180$ ns. For such a pulse period and spreading factor, $f_1 = 0.4$ GHz and $f_2 = 0.8$ GHz are good candidate frequencies which fit our target. [Fig. 4](#page--1-10) shows the non-chirped (a) and chirped (b) signals, the time duration of the latter is 25 ns, which is slightly less than double the original, with a chirping slope of 10 ns/GHz. As clearly noticed, the left and right portions of the chirped waveform have the same time behavior but different modulation frequencies. From [Fig. 5](#page--1-11) we can observe the better correlation between the received and transmitted waveforms when chirping is adopted, as a lower order of SOA nonlinearity is associated to chirped waveforms, while the amplifier input power is −21 dBm. Moreover, a higher signal-to-noise ratio is achieved, promising for an improvement in the bit error rate performance.

$$
XCTER = \frac{\text{Max}\{x_T(t) \star x_R(t)\}}{\int_{-T/2}^{+T/2} x_R^2(t)dt}
$$
(4)

The cross correlation to energy ratio (XCTER) between the electrically generated pulse $x_T(t)$ and received one $x_R(t)$ after the photodetector is expressed in Eq. [\(4\),](#page-1-4) this criteria is useful for evaluating the potential of chirping in mitigating SOA nonlinearities. XCTER has been studied versus the laser output power in [Fig. 6](#page--1-12) considering $I_{bias} = 100 \text{ mA}$ and $I_{bias} = 200 \text{ mA}$ with 10 ns/GHz chirping slope, the increasing response for different settings is due to the raise in the

Fig. 2. A schematic diagram for the experimental setup (Amp: Amplifier, MZM: Mach-Zehnder Modulator, SMF: Single Mode Fiber, SOA: Semiconductor Optical Amplifier, Att: Attenuator, PC: Polarization Controller, PD: Photo-detector).

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