Systems & Control Letters 61 (2012) 266-272

Contents lists available at SciVerse ScienceDirect

Systems & Control Letters

journal homepage: www.elsevier.com/locate/sysconle

Lyapunov formulation of the large-scale, ISS cyclic-small-gain theorem: The discrete-time case*

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ABSTRACT

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ARTICLE INFO

Article history: Received 10 August 2010 Received in revised form 20 May 2011 Accepted 1 November 2011 Available online 10 December 2011

Keywords: Dynamical networks Discrete-time Input-to-state stability (ISS) Small-gain Lyapunov functions

1. Introduction

The small-gain theorem has been widely recognized as one of the most important tools for stability analysis and control design in both linear and nonlinear systems. Truly nonlinear small-gain theorems, i.e., those allowing general nonlinear gains, started with [1,2] using input–output operator theory and [3] relying upon Sontag's seminal work on input-to-state stability (ISS) [4,5] and its various equivalent characterizations by Sontag and Wang [6]. Early applications of nonlinear small-gain to control systems with saturation and dynamic uncertainties appear in [7,8], respectively. The discrete-time counterpart of continuous-time ISS and its small-gain theorem were developed later in [9,10]. Considering the relation between ISS and Lyapunov functions, the Lyapunov formulations of ISS small-gain theorems for continuoustime systems and discrete-time systems were reported in [11,12].

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The Lyapunov-based small-gain results for hybrid systems have also been developed in [13–15].

This paper presents a Lyapunov formulation of the cyclic-small-gain theorem for dynamical networks

composed of discrete-time input-to-state stable (ISS) subsystems. ISS-Lyapunov functions for dynamical

networks satisfying the cyclic-small-gain condition are constructed from the ISS-Lyapunov functions of

Recently, some interesting extensions of the nonlinear smallgain theorem were obtained for general networks composed of ISS subsystems in [16-21] for both continuous-time and discrete-time systems. In [18], the interconnections in large-scale systems are formulated with nonlinear matrices and the small-gain condition was obtained in matrix form. In [20,21], the dynamical network with "max"-type interconnection was systematically studied and more general cyclic-small-gain criteria were developed for inputto-output stable (IOS) systems. The Lyapunov formulation of the matrix small-gain theorem for both continuous-time and discretetime systems was developed in [22]. The Lyapunov formulation of cyclic-small-gain for continuous-time dynamical networks was reported in our recent work [23]. It should be mentioned that the "max"-type interconnection can also be represented with the notion of monotone aggregation functions in the recent work [22]. Refs. [24-26] show the equivalence of small-gain conditions and the existence of an asymptotically stable discretetime comparison system induced by the interconnection gains of dynamical networks. In particular, Refs. [24,25] proved a vector small-gain theorem based on the relation between the vector gains in the general dynamical networks and the asymptotic stability of discrete-time systems with "MAX"-type system matrices.

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^{*} This research was supported under the Australian Research Councils Discovery funding scheme (project number: FF0455875), in part by a seed grant from NYU and POLY, and by NSF grant DMS-0906659.

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^{0167-6911/\$ –} see front matter s 2011 Elsevier B.V. All rights reserved. doi:10.1016/j.sysconle.2011.11.002

In view of the critical importance of discrete-time system theory in computer-aided control engineering applications [27], in this paper, we take a lead in generalizing the results of [23] to discrete-time dynamical networks. Interestingly, as we will see later, new phenomena arise; see Remarks 1 and 3 below. Also, some new technical lemmas are needed to achieve this generalization. Just like the continuous-time case, we can systematically construct a total ISS-Lyapunov function on the basis of ISS-Lyapunov functions of individual subsystems.

The rest of the paper is organized as follows. Section 2 gives some notations and definitions. Section 3 provides two ISS-Lyapunov formulations of the subsystems in the dynamical networks. Section 4 mainly studies the ISS-Lyapunov cyclic-small-gain for dynamical networks with subsystems formulated in dissipation form. In Section 5, we develop a counterpart of the results in Section 4 for dynamical networks with subsystems formulated in "gain margin" form. In Section 6, we employ an example to show the effectiveness of the main result. Section 7 presents some conclusions. The new technical lemmas are in Appendix.

2. Notations and definitions

Throughout the paper, we use |x| to denote the Euclidean norm of $x \in \mathbb{R}^n$ and x^T to denote the transpose of the vector x. We denote by \mathbb{Z}_+ the set of nonnegative integers.

A function $\gamma : \mathbb{R}_{\geq 0} \to \mathbb{R}_{\geq 0}$ is positive definite if $\gamma(s) > 0$ for all s > 0 and $\gamma(0) = 0$. $\gamma : \mathbb{R}_{\geq 0} \to \mathbb{R}_{\geq 0}$ is a class \mathcal{K} function if it is continuous, strictly increasing and $\gamma(0) = 0$; it is a class \mathcal{K}_{∞} function if it is a class \mathcal{K} function and also satisfies $\gamma(s) \to \infty$ as $s \to \infty$. For nonlinear functions γ_1 and γ_2 defined on $\mathbb{R}_{\geq 0}$, inequality $\gamma_1 \leq \gamma_2$ (or $\gamma_1 < \gamma_2$) represents $\gamma_1(s) \leq \gamma_2(s)$ (or $\gamma_1(s) < \gamma_2(s)$) for all s > 0. Id represents identity function.

We employ some notations in graph theory [28] to describe the interconnection of the dynamical network. The interconnection of the dynamical network can be represented with a directed graph by considering the subsystems as *vertices*. We use V_i to represent the *i*-th vertex (subsystem) in the graph. If the state of the *i*-th subsystem is an input of the *j*-th subsystem, then $\gamma_{ji} \neq 0$ and there is a *directed arc* (interaction) $(V_i V_j)$ from V_i to V_j ; otherwise, $\gamma_{ji} = 0$ and there is no arc from V_i to V_j . A path in the directed graph is any sequence of arcs where the final vertex of one is the initial vertex of the next one, denoted by $(V_i V_j \cdots V_k)$. If there exists a path leading from V_i to V_j , then V_j is *reachable* from V_i . Specifically, V_j is reachable from itself. The *reaching set RS* (V_j) is the set of the vertices other than the starting and ending vertices is called a *simple cycle*.

3. System formulation

The discrete-time dynamical network studied in this paper is composed of *N* subsystems in the following form

$$x_i(T+1) = f_i(x(T), u_i(T)), \quad i = 1, \dots, N;$$
(1)

where $x_i \in \mathbb{R}^{n_i}$, $x = [x_1^T, \ldots, x_N^T]^T$, $u_i \in \mathbb{R}^{n_{ui}}$ and $f_i : \mathbb{R}^{\sum_{j=1}^N n_j + n_{ui}} \rightarrow \mathbb{R}^{n_i}$ is continuous. *T* takes values in \mathbb{Z}_+ . Assume that $f_i(0, 0) = 0$, and the external input $u = [u_1^T, \ldots, u_N^T]^T$ is bounded. Denote $f(x, u) = [f_1^T(x, u_1), \ldots, f_N^T(x, u_N)]^T$.

It is assumed that for each x_i -subsystem (i = 1, ..., N), there exists a continuous ISS-Lyapunov function V_i such that the following properties hold.

(1) there exist $\underline{\alpha}_i, \overline{\alpha}_i \in \mathcal{K}_{\infty}$ such that

$$\underline{\alpha}_{i}\left(|x_{i}|\right) \leq V_{i}\left(x_{i}\right) \leq \overline{\alpha}_{i}\left(|x_{i}|\right), \quad \forall x_{i};$$

$$(2)$$

(2) there exist $\alpha_i \in \mathcal{K}_{\infty}, \sigma_{ij} \in \mathcal{K}_{\infty} \cup \{0\}$ and $\sigma_{ui} \in \mathcal{K} \cup \{0\}$ such that

$$V_{i}(f_{i}(x(T), u_{i}(T))) - V_{i}(x_{i}(T))) \\ \leq -\alpha_{i}(V_{i}(x_{i}(T))) \\ + \max_{j=1,\dots,N; j \neq i} \{\sigma_{ij}(V_{j}(x_{j}(T))), \sigma_{ui}(|u_{i}(T)|)\}.$$
(3)

Without loss of generality, we assume that $(Id - \alpha_i) \in \mathcal{K}$ as in [9]. Define

$$\hat{\gamma}_{ij} = \alpha_i^{-1} \circ (\mathrm{Id} - \rho_i)^{-1} \circ \sigma_{ij} \tag{4}$$

as the ISS gain from V_j to V_i , with ρ_i positive definite satisfying $(\text{Id} - \rho_i) \in \mathcal{K}_{\infty}$.

Correspondingly, we define

$$\hat{\gamma}_{ui} = \alpha_i^{-1} \circ (\mathrm{Id} - \rho_i)^{-1} \circ \sigma_{ui}.$$
(5)

Remark 1. There are two Lyapunov formulations of discrete-time ISS systems: "gain margin" form and dissipation form (refer to [9,29]). The dissipation form is in the form of (3), while the "gain margin" form can be described as there exist a continuous and positive definite α'_i and γ'_{ij} , $\gamma'_{ui} \in \mathcal{K} \cup \{0\}$ such that

$$V_{i}(x_{i}(T)) \geq \max \left\{ \gamma_{ij}'\left(V_{j}\left(x_{j}(T)\right)\right), \gamma_{ui}'\left(|u_{i}(T)|\right) \right\} \Rightarrow V_{i}\left(f_{i}\left(x\left(T\right), u_{i}\left(T\right)\right)\right) - V_{i}\left(x_{i}\left(T\right)\right) \leq -\alpha_{i}'\left(V_{i}\left(x_{i}\left(T\right)\right)\right).$$
(6)

However, different from continuous-time systems, the trajectory of the discrete-time system may "jump" out of the region determined by the "gain margin", which means that γ'_{ij} and γ'_{ui} may not be the real nonlinear gains from V_j 's and u_i to V_i . However, as done in [12,13], to prove small-gain results the "gain margin" form formulation (6) should be consolidated with

$$V_{i}(x_{i}(T)) \leq \max \left\{ \gamma_{ij}'\left(V_{j}\left(x_{j}(T)\right)\right), \gamma_{ui}'\left(|u_{i}(T)|\right) \right\} \\ \Rightarrow V_{i}\left(f_{i}\left(x\left(T\right), u_{i}\left(T\right)\right)\right) \\ \leq \left(\mathrm{Id} - \delta_{i}'\right)\left(\max \left\{\gamma_{ij}'\left(V_{j}\left(x_{j}(T)\right)\right), \gamma_{ui}'\left(|u_{i}(T)|\right)\right\}\right)$$

$$(7)$$

where δ'_i is continuous and positive definite satisfying $(\text{Id} - \delta'_i) \in \mathcal{K}_{\infty}$. \Box

Combining (6) and (7), the "gain margin" formulation for smallgain analysis is obtained by replacing property (2) with

(2') there exist $\gamma_{ij} \in \mathcal{K}_{\infty} \cup \{0\}$ and $\gamma_{ui} \in \mathcal{K}_{\infty} \cup \{0\}$ such that

$$\leq (\mathrm{Id} - \delta_i) \begin{pmatrix} \max_{j \in \{1, \dots, N\} \setminus \{i\}} \begin{cases} \gamma_{ij} \left(V_j \left(x_j(T) \right) \right), \\ V_i \left(x_i(T) \right), \\ \gamma_{ui} \left(|u_i(T)| \right) \end{cases} \end{pmatrix}$$

$$(8)$$

where δ_i is continuous and positive definite satisfying $(Id - \delta_i) \in \mathcal{K}_{\infty}$.

For the sake of generality of our main result, we will first consider the dissipation formulation, and then extend the result to the "gain margin" form following a similar idea. Download English Version:

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