# Mutually orthogonal matrices from division algebras 

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## A R T I C L E I N F O

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#### Abstract

Matrices $A$ and $B$ in $M_{n}(\mathbb{C})$ are said to be mutually orthogonal if $A B^{*}+B A^{*}=0$, where * denotes the conjugate transpose. We study cardinalities of certain $\mathbb{R}$-linearly independent families of matrices arising from matrix embeddings of a division algebra of index $m$ with center a number field $Z$, satisfying the property that matrices from different families are mutually orthogonal. The question is of importance in the context of coding for certain wireless channels, where the cardinalities of such sets is connected to the maximum code rate consistent with low decoding complexity. It follows from our results that the maximum code rate for the codes we consider is severely limited.


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## 1. Introduction

This paper deals with a question that arises from certain coding and decoding issues in wireless communication. Let $D$ be a division algebra of index $m$, and center a number field $Z$. Suppose that we have an embedding $\phi: D \rightarrow M_{n}(\mathbb{C})$ for some $n$. Thus, $\phi$ is a (necessarily injective) ring homomorphism, which by definition takes $1_{D}$ to the identity matrix. We will work exclusively with the embedded forms $\phi(D)$ and $\phi(Z)$, and by abuse of notation, will continue to write $D$ and $Z$ respectively for $\phi(D)$ and $\phi(Z)$. We will call two matrices $A$ and $B$ in $M_{n}(\mathbb{C})$ mutually orthogonal if $A B^{*}+B A^{*}=0$, where * denotes the conjugate transpose. Suppose $\Gamma_{1}, \Gamma_{2}, \Gamma_{3}$, and $\Gamma_{4}$ are four (nonempty) families of matrices in $D$ such that any two matrices from distinct families are mutually orthogonal and such that the matrices in $\Gamma_{1} \cup \Gamma_{2} \cup \Gamma_{3} \cup \Gamma_{4}$ are $\mathbb{R}$-linearly independent. Assume that $\left|\Gamma_{1}\right|=\left|\Gamma_{2}\right|=\left|\Gamma_{3}\right|=\left|\Gamma_{4}\right|=k$. The question we study is the following: What is the maximum value of $k$ ? Under the assumption that $Z=Z^{*}$, which arises quite naturally in the application to wireless communication, we show that this maximum is $m d / 2$, where $d$ is the degree of the minimal polynomial of $\alpha$ as a matrix in $M_{n}(\mathbb{C}), \alpha$ a generator of $Z$ over $\mathbb{Q}$. Here, $m$ is necessarily even, and we identify $\mathbb{Q}$ with its image $\phi: q \mapsto \operatorname{diag}(q, \ldots, q)$ in $M_{n}(\mathbb{C})$. We give examples to show that this maximum is actually attained.

[^0]Our main theorem is the following:
Theorem 1. With notation and assumptions as above:
(1) The index $m$ of $D$ is even.
(2) We have $k \leq \frac{m t}{2}$, where $t$ is the maximum number of $\mathbb{R}$-linearly independent Hermitian matrices in $Z$. Further, $t \leq d$, where $d$ is the degree of the minimal polynomial (as a matrix in $M_{n}(\mathbb{C})$ ) of any $\alpha \in Z$ such that $Z=\mathbb{Q}(\alpha)$. Thus, $k \leq \frac{m d}{2}$.
(3) $m d \leq n$, so $k \leq \frac{n}{2}$.

We begin our considerations in the next section, but we first describe briefly how this question arises. In the field of multiple-antenna communication, division algebras embedded in $M_{n}(\mathbb{C})$ form natural candidates for constructing space-time block codes, which for our purpose are matrices $X(\underline{s})$ arising from the embedded division algebra, whose entries depend linearly on a $2 l$-tuple $\underline{s}=\left(s_{1}, \ldots, s_{l}, s_{1}^{*}, \ldots, s_{l}^{*}\right), l \leq n^{2}$. Here, the $s_{i}$ take values in a finite subset $\mathcal{S}$ of the nonzero complex numbers. The $l$-tuple $\left(s_{1}, \ldots, s_{l}\right)$ carries the message to be transmitted, and the matrix size $n$ signifies both the number of antennas used as also the number of uses of the transmission channel in one block of transmission. (See [8], [9] for instance. Note that these references mainly focus on the situation where $X(\underline{s})$ depends only on $s_{1}, \ldots, s_{l}$, but we can just as easily allow the more general case where $X(\underline{s})$ also depends on the complex conjugates $s_{i}^{*}$.) Writing each $s_{i}$ as $a_{2 i-1}+\beta a_{2 i}$, where $a_{2 i-1}$ and $a_{2 i}$ are real and $\beta$ is non-real, the code matrices can be written in the form

$$
\begin{equation*}
X=X\left(a_{1}, \ldots, a_{2 l}\right)=\sum_{i=1}^{2 l} a_{i} A_{i} \tag{1}
\end{equation*}
$$

where the $A_{i}$ are fixed $n \times n$ complex matrices. The message is now carried by the real $2 l$-tuple ( $a_{1}, a_{2}, \ldots$ ), where the $a_{i}$ come from the set $\mathcal{R}(\mathcal{S})$ defined as $\{x \in \mathbb{C} \mid x+\beta y \in \mathcal{S}$, for some $y \in \mathbb{C}\}$ unioned with $\{y \in \mathbb{C} \mid x+\beta y \in \mathcal{S}$, for some $x \in \mathbb{C}\}$. Note that the $A_{i}$ must be $\mathbb{R}$-linearly independent, else, we could write some $A_{i}$ as an $\mathbb{R}$-linear combination of the remaining $A_{j}$ in the right side of Equation (1), and as a result, we would effectively be transmitting fewer than $2 l$ real symbols and hence less information in each matrix $X$.

Typically, the division algebra $D$ from which the matrices $X(\underline{s})$ arise is taken to be an $Z$-central division algebra, where $Z$ is one of $\mathbb{Q}, \mathbb{Q}(\imath)$, or $\mathbb{Q}(\omega)$, where $\omega$ is a primitive 3rd root of unity, and $S$ is taken to be a subset of the nonzero elements of $Z$. When $Z=\mathbb{Q}(\imath), \beta$ above is taken to be $\imath$, and when $Z=\mathbb{Q}(\omega)$, $\beta$ above is taken to be $\omega$. In such situations, and under the assumption that $X(\underline{s}) \in D$ for all $s \in Z$, as is the situation in practice, it is easy to see that the $A_{i}$ themselves are also in $D$. (Of course, when $Z=\mathbb{Q}$, there are no imaginary parts to the $s_{i}$, instead, for uniformity of notation, we will tacitly assume that in this case, $\underline{s}$ is really a $2 l$-tuple ( $a_{1}, \ldots, a_{2 l}$ ), with $2 l \leq 2 n^{2}$.)

When some standard lattice-based decoding procedures are employed, the decoding process has worst-case decoding complexity of the order $\mathcal{O}\left(\mid \mathcal{R}(\mathcal{S} \mid)^{2 l}\right)$, which, especially when the code is "full-rate" (i.e., $l=n^{2}$ ), is prohibitively high. It is of interest to reduce the exponent of $|\mathcal{R}(\mathcal{S})|$ in the complexity, by enabling the $a_{i}$ to be decoded in parallel groups. If this can be accomplished, and if say $k$ is the maximum size of the groups, then the decoding complexity drops to $\mid \mathcal{R}(\mathcal{S} \mid)^{k}$. Suppose that the symbols $a_{1}, \ldots, a_{2 l}$ can be decoded in parallel in groups $\Gamma_{1}, \ldots, \Gamma_{g}$, with $\Gamma_{i}$ (after reindexing $a_{1}, \ldots, a_{2 l}$ ) containing the symbols $a_{i, 1}, a_{i, 2}, \ldots$ We may rewrite Equation (1) as

$$
\begin{equation*}
X=\sum_{i=1}^{g} \sum_{u} a_{i, u} A_{i, u} \tag{2}
\end{equation*}
$$

where the $A_{i}$ are correspondingly partitioned and reindexed. An analysis of the decoding process shows that decoding can occur in such parallel groups if and only if each of the corresponding matrices $A_{i, u}, u=1,2, \ldots$,

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